Detection of Heart Rate Variability From a Wearable Differential ECG Device

Jure Slak*, Gregor Kosec**

*student at Jožef Stefan Insitute, Department of Communication Systems, Ljubljana, Slovenia **Jožef Stefan Insitute, Department of Communication Systems, Ljubljana, Slovenia jure.slak@student.fmf.uni-lj.si, gregor.kosec@ijs.si

Abstract - The precise heart rate variability is extracted from an ECG signal measured by a wearable sensor that constantly records the heart activity of an active subject for several days. Due to the limited resources of the wearable ECG device the signal can only be sampled at relatively low, approximately 120 Hz, frequency. Besides low sampling rate the signal from a wearable sensor is also burdened with much more noise than the standard 12-channel ambulatory ECG, mostly due to the design of the device, i.e. the electrodes are positioned relatively close to each other, and the fact that the subject is active during the measurements. To extract heart rate variability with 1 ms precision, i.e. 10 times more accurate than the sample rate of the measured signal, a two-step algorithm is proposed. In first step an approximate global search is performed, roughly determining the point of interest, followed by a local search based on the Moving Least Squares approximation to refine the result. The methodology is evaluated in terms of accuracy, noise sensitivity, and computational complexity. All tests are performed on simulated as well as measured data. It is demonstrated that the proposed algorithm provides accurate results at a low computational cost and it is robust enough for practical application.

I. INTRODUCTION

It is well known that the morphology of ECG signals changes from beat to beat as a consequence of physical activity, sensations, emotions, breathing, etc. of the subject [1]. The most straightforward measure of these changes is the heart rate variability (HRV), i.e. small variations of beat duration. HRV characterizes the timings of hearth cells repolarization and depolarization processes. The HRV is typically determined by measuring the intervals between two consecutive R-waves (RRI) or intervals between R and T waves (RTI). Several vital signals can be identified from the HRV and therefore it is often used as a health status indicator in different fields of medicine, e.g. neurology [11], cardiac surgery [5], heart transplantation [3] and many more. Typical HRV values of healthy subjects are approximately 40 ms for RRI and 2 ms for RTI [2] (see Figure 1). Therefore it is important to detect considered waves with at least 1 ms accuracy for practical use [6]. This paper deals with the detection of HRV in ECG signal provided by a Wearable ECG Device (WECGD) that is paired with a personal digital assistant (PDA) via Bluetooth Smart protocol [4]. The WECGD, due to the hardware limitations, only measures the signal, while the PDA takes care of data visualization, basic analysis and transmission of the data to a more powerful server for further analyses. In contrast to a standard ambulatory 12-channel ECG measurement, where trained personnel prepare and supervise the measurement of subject at rest, the WECGD works on a single channel, the subject is active and since the WECGD is often placed by an untrained user its orientation might be random, resulting in additional decrease of signal quality. In order to maintain several days of battery autonomy The WECGD also records the heart rate with significantly lower frequencies and resolution in comparison to ambulatory measurements. In this paper we analyse a possible local, i.e. only short history of measurement data is required, algorithm for detection of heart rate variability with 1 ms precision of a signal recorded with 120 Hz.

The rest of the paper is organized as follows. In section II a detection algorithm is presented, in section III results are discussed, and in section IV conclusions are presented.

II. DETECTION METHOD

In order to evaluate the HRV, the *characteristic point* of each heart beat has to be detected in the signal, which is provided as values of electric potential sampled with a frequency 120 Hz. Since the HRV is computed from differences of consequent characteristic points the choice of the characteristic point does not play any role, as long as it is the same in every beat. In this work we choose to characterise the beat with a minimal first derivative, in another words, we seek the points in the signal with the most violent drop in electric potential (Figure 1) that occurs between R and S peaks.

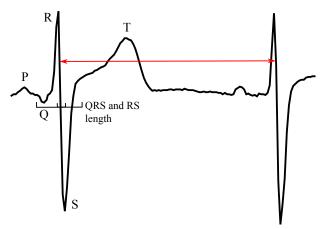
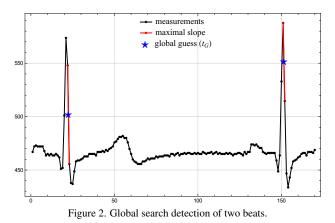


Figure 1. Beat to beat time between two characteristic points.

The detection method is separated in two stages, namely global and local. The goal of the global method is to approximately detect the characteristic point, while the local method serves as a fine precision detection, enabling us to detect HRV with much higher accuracy.

A. Coarse global search

In the first step the algorithm finds a minimal first derivative of a given signal on a sample rate accuracy, i.e. $1/\nu$. The global search method is next to trivial. The algorithm simply travels along the signal, calculating the discrete derivative and storing the position of minimal values found so far. Since the points are sampled equidistantly, minimizing $\frac{\Delta y}{\Delta t}$ is equal to minimizing Δy . The middle of the interval where the largest drop was detected is taken as the global guess t_G . The results of the global search are presented in Figure 2.



B. Fine local search

The global search provides only coarse, limited to sample points, positions of the characteristic points. To push accuracy beyond $1/\nu$, the signal has to be represented also in between the sample points. A monomial approximation function based on a Moving Least Squares approach [10] is introduced for that purpose.

The value of the electrical potential at arbitrary time t_0 is approximated. Denote the vector of n known values near t_0 by f (called *support*), and the times at which they were measured by t. The approximation \hat{f} of f is introduced as a linear combination of m in general arbitrary basis functions $(b_j)_{j=1}^m$, however in this work only monomials are considered.

$$\hat{f} = \sum_{j=1}^{m} \alpha_j b_j \tag{1}$$

The most widely used approach to solve above problem and find the appropriate \hat{f} is to minimize the weighted 2-norm of the error, also known as the Weighted Least Squares (WLS) method.

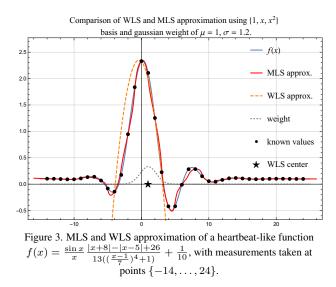
$$\|\boldsymbol{f} - \hat{f}(\boldsymbol{t})\|_{w}^{2} = \sum_{i=1}^{n} (f_{i} - \hat{f}(t_{i}))^{2} w(t_{i}), \qquad (2)$$

In the above equation w is a nonnegative weight function.

The only unknown quantities are the m coefficients α of the linear combination, which can be expressed as

a solution of an overdetermined linear system $WB\alpha = Wf$, where W is the $n \times n$ diagonal weight matrix, $W_{ii} = \sqrt{w(t_i)}$ and B is the $n \times m$ collocation matrix, $B_{ij} = b_j(t_i)$. There are different approaches towards finding the solution. The fastest and also the least stable and accurate is to solve the Normal System $B^T W^T W B\alpha = B^T W^T W f$, a more expensive but also more stable is via QR decomposition, and finally the most expensive and also the most stable is via SVD decomposition [9]. The resulting vector α is then used to calculate $\hat{f}(t)$ for any given t. The derivatives are approximated simply by differentiating the approximating function, $\hat{f}' = \sum_{j=1}^m \alpha_j b'_j$.

The WLS approximation weights the influence of support points using the weight function w. Usually, such a weight is chosen that points closest to t_0 are more important in the norm in comparison to the nodes far away. Naturally such approximation is valid only as long as the evaluation point is close to the t_0 . A more general approach is a Moving Least Square (MLS) approximation, where coefficients α are not spatially independent any more, but are recomputed for each evaluation point. Naturally, such an approach is way more expensive, but also more precise. A comparison of both methods, i.e. WLS and MLS, is shown in Figure 3.



The task of finding the minimal value of the first derivative is equivalent to the task of finding the zero of the second derivative. This zero will be our local approximation t_L of the beat time, $\hat{f}''(t_L) = 0$. Therefore an approximation function with a non constant second derivative, i.e. approximation function with a minimal 3rd order monomial basis, is constructed. The most straightforward approach to find its root is the simple bisection. Bisection requires initial low and high bounds that can be estimated from characteristic point t_G provided by the global method. Using the fact that the QRS intervals last approximately $\Delta t_{QRS} = 0.1 \text{ s}$ [8, 7], we can seek for the root of the second derivative on an interval $[t_G - \Delta t_{QRS}/2, t_G + \Delta t_{QRS}/2]$, and at given sample rate this translates to the search interval of two sample points away from t_G in each direction.

C. HRV calculation and error estimation

Given sampled heartbeat, the fine local search described in section B produces the vector of $\ell + 1$ detected times $t_L := (t_{L,i})_{i=1}^{\ell+1}$ of the RS slopes. Their successive differences represent a vector \hat{r} of detected beat to beat times, the durations of RR intervals.

$$\hat{\boldsymbol{r}} = (\hat{r}_i)_{i=1}^{\ell}, \quad \hat{r}_i = t_{L,i+1} - t_{L,i}$$

Let r be the vector of (usually unknown) actual beat to beat times. Then the heart rate variability (HRV) h is defined as

$$h := \operatorname{std}(\boldsymbol{r}) = \sqrt{\frac{1}{\ell} \sum_{i=1}^{\ell} (r_i - \bar{r})^2},$$

where \bar{r} stands for the average beat to beat time, $\bar{r} = \sum_{i=1}^{\ell} r_i / \ell$. The HRV estimation \hat{h} is calculated as the standard deviation of the detected times \hat{r} .

In the following analyses the actual vector \boldsymbol{r} will be known, since the synthesized heartbeat will be analysed. The most obvious error measures are the absolute error of HRV, $e_h = |\hat{h} - h|$ and the absolute error of the average heart beat $e_{\bar{r}} = |\bar{r} - \bar{r}|$. Using the vector of errors $\boldsymbol{e} = |\boldsymbol{r} - \hat{\boldsymbol{r}}|$ the average error $e_a = \sum e_i/\ell$ and the maximal error $e_M = \max(\boldsymbol{e})$ can be assessed.

III. RESULTS AND DISCUSSION

A. Approximation set-up

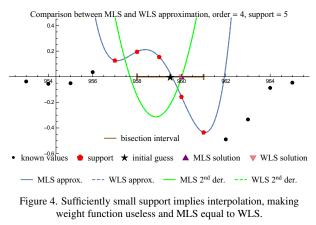
In first step the approximation free parameters, i.e. weight function, support size, and number of basis functions, have to be assessed. A single heartbeat is extracted and approximated with all possible combinations of basis functions with orders from 2 to 10 and symmetric supports of sizes from 3 to 15 using both WLS and MLS. An global algorithm described in section A was used to produce the initial guesses. For demonstration four sample cases are presented. The weight function was the same in all four cases, a Gaussian distribution with $\mu = t_G$ and $\sigma = m/4$, which makes sure that all support points are taken into account, but the central ones are more important.

The simplest case is when the support size is the same as number of basis functions resulting in an interpolation. In this case, the weight function is not important, making WLS and MLS entirely equivalent as seen in Figure 4.

In case of small support and small order of monomial basis WLS performs worse than MLS and the approaches differ significantly. However, as we increase the order of the polynomial basis the difference within the bisection interval becomes negligible. This transition can be observed in Figures 5 and 6.

As predicted, the support size is important. Both methods perform badly when too many surrounding measurements are taken into account while still using a low order polynomial approximation. Note that in Figure 7 the initial guess is barely improved and the beat shape is skewed away from the RS drop.

The conclusion is, that for our purposes MLS approximation is unnecessary, as WLS provides good enough results, when used appropriately. Further analysis to determine the best choice of parameters m and n is presented in section C.



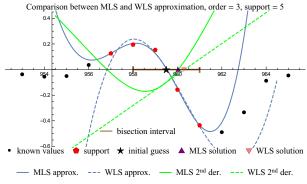


Figure 5. MLS and WLS differ when approximating with a low order polynomial.



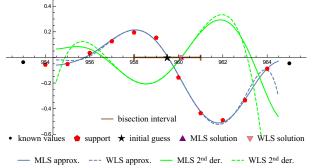


Figure 6. MLS and WLS match when approximating with a high order polynomial.

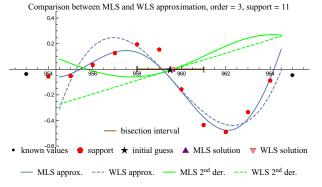


Figure 7. Expected bad behaviour with too many support points and a low order approximation.

B. Computational complexity

The presented algorithm is a streaming algorithm, requiring a buffer to store the current beat in the signal. Let b be the number of measurements per beat, stored in a buffer of length b. The global part of the algorithm makes O(b) operations, being a simple local search. The local part is more expensive. First a $n \times m$ matrix is constructed in O(mn) and the right hand side vector is copied from the buffer. The system is then solved using SVD decomposition in $O(mn^2 + n^3)$. Note that as m = O(n), this step takes $O(n^3)$. The minimal first derivative is found using bisection. To achieve tolerance ε , $\lceil \log_2(1/\varepsilon) \rceil$ function evaluations are needed, each costing O(m) operations. Total time complexity is there-fore equal to $O(b + n^3 + m \log(1/\varepsilon))$. Note, that using MLS would require $O(n^3)$ for each function evaluation, resulting in a significantly worse time complexity of $O(b + n^3 \log_2(1/\varepsilon)))$. The calculation of average and variance is done later, after the wanted amount of signal has already been analysed.

In practice the algorithm executes very fast, using typical values b = 150, m = 6, n = 11 and $\varepsilon = 10^{-10}$ it runs approximately 0.27 s to analyse 1000 heartbeats ($\approx 10^5$ data points). The algorithm was compiled from C++ source code using g++5.3.0 with -02 flag and run on an Intel (R) Core (TM) i7-4700MQ processor.

C. Simulated heartbeat with known variability

The first set of tests for the presented method was performed using a simulated heartbeat. A single real heartbeat was taken and then replicated thousand times, each time shifted by a random offset T, distributed normally around zero, $T \sim \mathcal{N}(0, \sigma^2)$, with $\sigma = \frac{1}{2\nu} = \frac{1}{2}\Delta t$. This means that a decent amount of measurements will be more than Δt apart, a difference that must be detected by global search for method to work. However, around half of the measurements are less than Δt apart, forming suitable ground for testing the precision of the local search. At given sample frequency, σ equals 4.167 ms.

The generated beat to beat times, coarsely detected and finely detected times are presented in Figure 8.

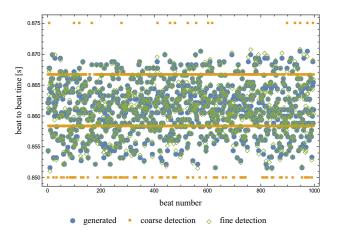


Figure 8. Generated beat to beat times and their global detection.

Beat to beat time precision is significantly improved by the local search. As seen in Figure 9, the distributions of generated and detected heart beats match very well.

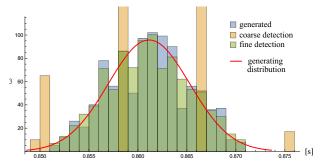


Figure 9. Generated RRI times and their global detection. The middle two coarse detection columns continue off the chart, but are not shown completely for clarity.

Results of RRI and HRV detection by global and local search are presented in Table 1. The generated times were taken as precise and the algorithm was run to produce global and local approximations. Then the average RRI time and HRV were calculated for each data set separately. The average RRI time is estimated very well with both methods, but the precision of the global method is not satisfactory when measuring heart rate variability. The precision is significantly improved using the local search.

A chart showing the average error of the detected times is shown in Figure 10. The red values indicate invalid region with more basis functions than support points. Although the under-determined system could be solved with SVD, in this paper we do not consider that option. Increasing the order of approximations improves results in the beginning, however ultimately a stability limit is reached where approximation system becomes unstable, i.e. the condition number of the matrix becomes critical, resulting in decrease of accuracy to the level of a coarse detection approach. The best results are obtained with order of basis 10–12 and 13–15 support values. From Figure 10 we can also conclude that using MLS is not beneficial enough to justify its computational cost.

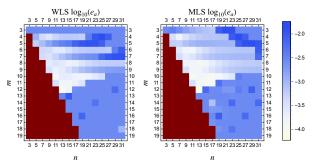


Figure 10. Comparison of WLS and MLS errors using different orders and support sizes.

TABLE 1. RESULTS AND ERRORS OF THE RRI AND HRV DETECTION.

quantity [s]	generated	coarse	fine
\bar{r}	0.861136	0.861139	0.861136
$e_{\bar{r}}$	0	$3.34 \cdot 10^{-6}$	$2.83 \cdot 10^{-8}$
h	0.004102	0.005324	0.004137
e_h	0	0.001222	$3.52 \cdot 10^{-5}$
e_a	0	0.002969	0.000263
e_M	0	0.007778	0.000829

D. Noise analysis

The method presented in this paper relies heavily on the values of the derivative. Normally, derivatives are very sensitive to noise. To analyse noise sensitivity of the algorithm a real heart beat signal was taken and normalized by subtracting the average and then divided by maximal absolute value, transforming measurements onto the interval [-1, 1]. A uniform noise of level p was then applied to every measurement. Specifically, a uniformly distributed random number on the interval [-p, p], for fixed $p \in [0, 1]$ was added to every measurement point.

Example of the noised and normalized beat and the behaviour of the algorithm is shown in Figure 11.

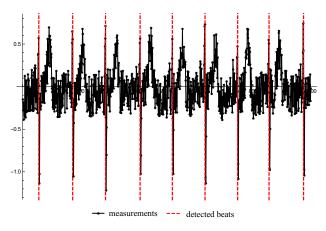


Figure 11. Detecting heart beat in a 25 % uniformly noised signal.

An approximate calculation of the critical noise level can be made. Let ν be the sample frequency and p the noise level. The critical noise level is such, that the maximal noise derivative is comparable to the maximal derivative of the signal. The maximal noise derivative is achieved when to subsequent measurements take extreme values of p and -p, resulting in drop of magnitude 2p. In the actual heartbeat, the maximal drop appears during R and S peaks. Taking maximal number of nodes in QRS complex to be $\lceil 0.1\nu \rceil$ as before, and with respect to Figure 1 the RS drop contains at most one third of them, we approximate the maximal drop between two subsequent measurements as the total RS drop divided by number of nodes included. Taking into account that the RS drop equals 2 after normalization, we obtain

$$\Delta y_{\max} = \frac{2}{\frac{1}{3} \lceil 0.1\nu \rceil} \approx 60 \frac{1}{\nu}.$$

The critical noise level is such, that both noise levels are approximately equal

$$2p_{\rm crit} \approx 60 \frac{1}{\nu}.$$

At sample frequency of 120 Hz this yields

$$p_{\rm crit} \approx 0.25.$$

The algorithm was tested as described above for noise levels $p \in [0, 1]$ with a step of 1 %. Results are presented in Figure 12. Note that the rise in error corresponds nicely with predicted critical noise.

A little bit more can be achieved using simple statistical analysis. When noise levels are around critical, a few false beats may be recognized among the actual ones. The RRI time before and after the false beat will deviate greatly from the average RRI time. Therefore, we can choose to ignore extreme values, as they are very likely to be wrong. Average and standard deviation are both sensitive to extreme data points, however the median is not, and while extreme values occur rarely, the median should give a good sense of what the "correct" beat to beat time is. To detect the extreme values a Median Absolute Deviation (MAD) was chosen as the simplest and robust enough approach. In this method, the median of differences from the median is taken as the estimate of data variability. All values that differ more that five times MAD from the median are ignored. The results of applying this technique are also shown in Figure 12.

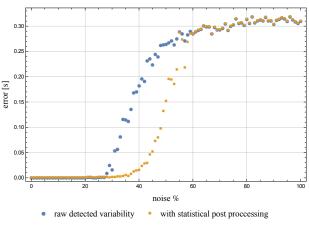


Figure 12. Variability detection at different noise levels.

E. Actual heartbeat analysis

The presented method was tested on sample inputs from the wearable sensor. Detected beats are presented in Figure 13. Average beat to beat time of the subject was $\hat{r} = 0.888264 \,\mathrm{s}$ and heart rate variability equals $\hat{h} = 0.282443 \,\mathrm{s}$.

IV. CONCLUSION

In this paper a simple MLS based algorithm for accurate detection of HRV from a low sample rate ECG signal, typically provided by a wearable sensor, is demonstrated. The algorithm is formulated in a fairly general way. Most of the approximation parameters can be easily changed. However, to keep the analyses within the reasonable limits, only the order of monomial basis and the support size are varied to find the optimal set-up. It is demonstrated that increasing order of basis as well as support size improves results up to roughly m = 12 and/or n = 20, at which point the system matrix becomes ill-conditioned. Based on the presented analyses a basis of 10th order supported with 15 nodes is claimed to be the optimal set-up. It is also demonstrated that using a way more computationally expensive MLS, in comparison to WLS, does not improve accuracy enough to justify it.

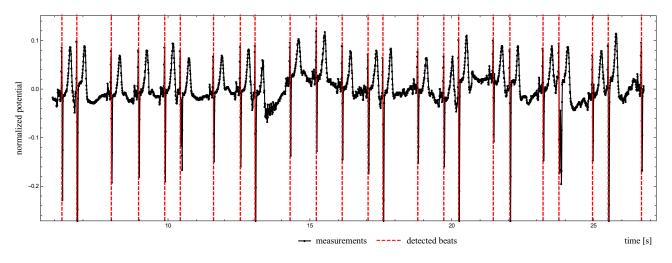


Figure 13. Detection of heart beat from the input measured by wearable sensor.

Based on results from approximation analyses a fast two-stage streaming algorithm for HRV detection is developed. The algorithm is tested on synthetic as well as actual data, achieving good performance. It is demonstrated that the detected beat times of a simulated heartbeat differ from the actual ones with an average absolute error of 0.263 ms at sample frequency of 120 Hz. In other words the detection accuracy is roughly ten times more accurate in comparison with a coarse one. The method is also stable up to approximately 25 % noise and computationally extremely effective, successfully processing 1000 heartbeats in approximately 0.27 s on a standard laptop computer.

In future work we will focus on full analysis, including different basis and weight functions, as well as non-linear MLS to fit also the basis shape parameters.

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